

# ECEN 667

## Power System Stability

### Lecture 12: Exciter Models

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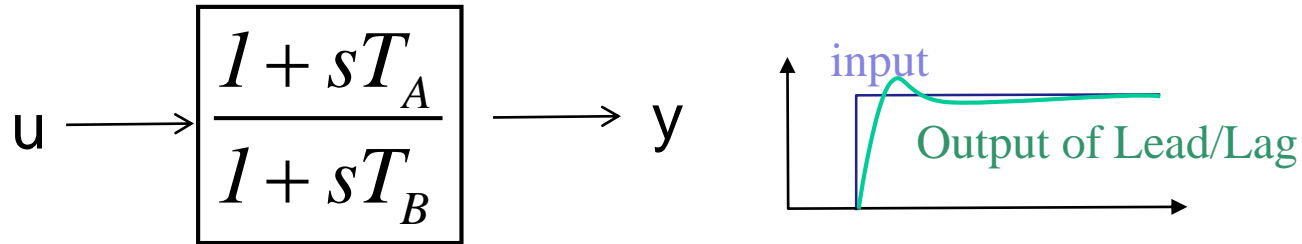
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UNIVERSITY

# Announcements



- Read Chapter 4
- Homework 4 is posted; it should be done before the first exam but need not be turned in
- Midterm exam is on Tuesday Oct 17 in class; closed book, closed notes, one 8.5 by 11 inch hand written notesheet allowed; calculators allowed

# Lead-Lag Block



- In exciters such as the EXDC1 the lead-lag block is used to model time constants inherent in the exciter; the values are often zero (or equivalently equal)
- In steady-state the input is equal to the output
- To get equations write in form with  $\beta_0=1/T_B$ ,  $\beta_1=T_A/T_B$ ,  $\alpha_0=1/T_B$

$$\frac{1 + sT_A}{1 + sT_B} = \frac{\frac{1}{T_B} + s\frac{T_A}{T_B}}{\frac{1}{T_B} + s}$$

# Lead-Lag Block

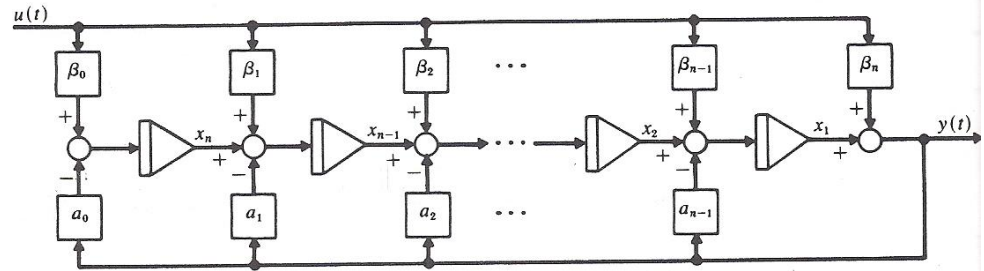


- The equations are with

$$\beta_0 = 1/T_B, \quad \beta_1 = T_A/T_B,$$

$$\alpha_0 = 1/T_B$$

then



$$\frac{dx}{dt} = \beta_0 u - \alpha_0 y = \frac{1}{T_B} (u - y)$$

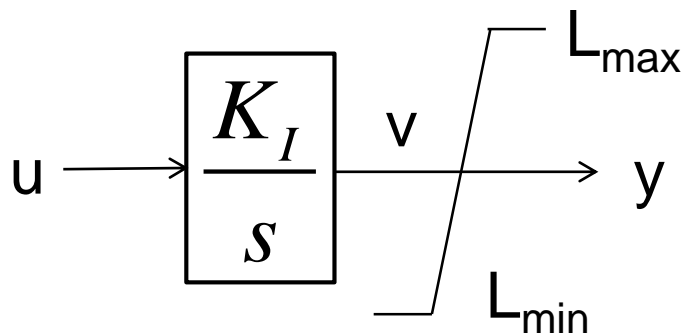
$$y = x + \beta_1 u = x + \frac{T_A}{T_B} u$$

The steady-state requirement that  $u = y$  is readily apparent

# Limits: Windup versus Nonwindup



- When there is integration, how limits are enforced can have a major impact on simulation results
- Two major flavors: windup and non-windup
- Windup limit for an integrator block



$$\frac{dv}{dt} = K_I u$$

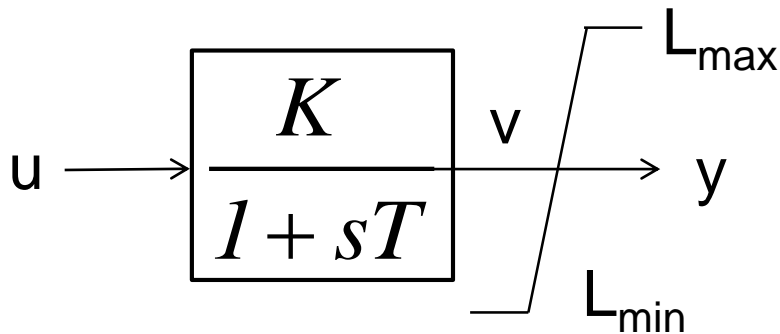
If  $L_{\min} \leq v \leq L_{\max}$  then  $y = v$   
else if  $v < L_{\min}$  then  $y = L_{\min}$ ,  
else if  $v > L_{\max}$  then  $y = L_{\max}$

The value of  $v$  is NOT limited, so its value can "windup" beyond the limits, delaying backing off of the limit

# Limits on First Order Lag



- Windup and non-windup limits are handled in a similar manner for a first order lag



$$\frac{dv}{dt} = \frac{1}{T}(Ku - v)$$

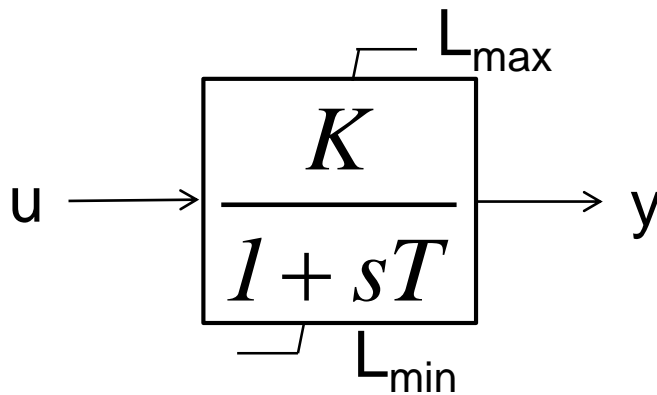
If  $L_{\min} \leq v \leq L_{\max}$  then  $y = v$   
else if  $v < L_{\min}$  then  $y = L_{\min}$ ,  
else if  $v > L_{\max}$  then  $y = L_{\max}$

Again the value of  $v$  is NOT limited, so its value can "windup" beyond the limits, delaying backing off of the limit

# Non-Windup Limit First Order Lag



- With a non-windup limit, the value of  $y$  is prevented from exceeding its limit



$$\frac{dy}{dt} = \frac{1}{T} (Ku - y)$$

(except as indicated below)

$$\text{If } L_{\min} \leq y \leq L_{\max} \text{ then normal } \frac{dy}{dt} = \frac{1}{T} (Ku - y)$$

$$\text{If } y \geq L_{\max} \text{ then } y=L_{\max} \text{ and if } u > 0 \text{ then } \frac{dy}{dt} = 0$$

$$\text{If } y \leq L_{\min} \text{ then } y=L_{\min} \text{ and if } u < 0 \text{ then } \frac{dy}{dt} = 0$$

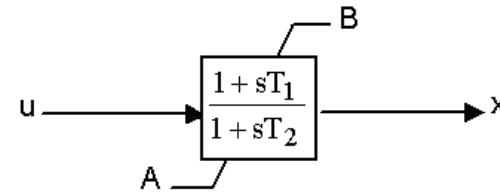
# Lead-Lag Non-Windup Limits



- There is not a unique way to implement non-windup limits for a lead-lag.

This is the one from  
IEEE 421.5-1995

(Figure E.6)



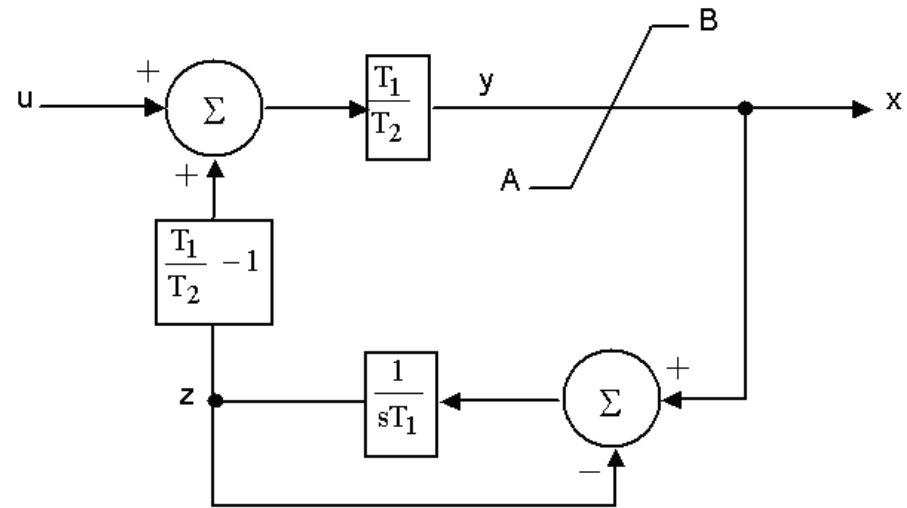
(a) Model

$$T_2 > T_1, T_1 > 0, T_2 > 0$$

If  $y > B$ , then  $x = B$

If  $y < A$ , then  $x = A$

If  $B \geq y \geq A$ , then  $x = y$



(b) Implementation

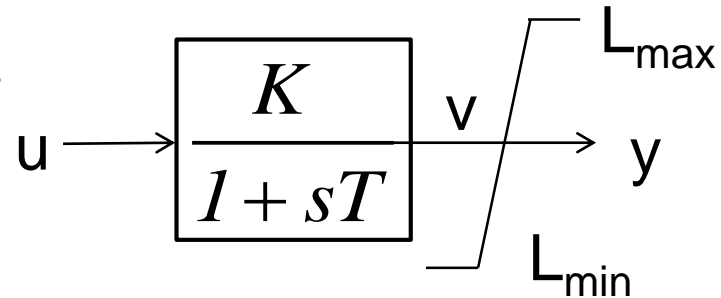


# Ignored States



- When integrating block diagrams often states are ignored, such as a measurement delay with  $T_R=0$
- In this case the differential equations just become algebraic constraints

- Example: For block at right, as  $T \rightarrow 0$ ,  $v=Ku$



- With lead-lag it is quite common for  $T_A=T_B$ , resulting in the block being ignored

# IEEE T1 Example



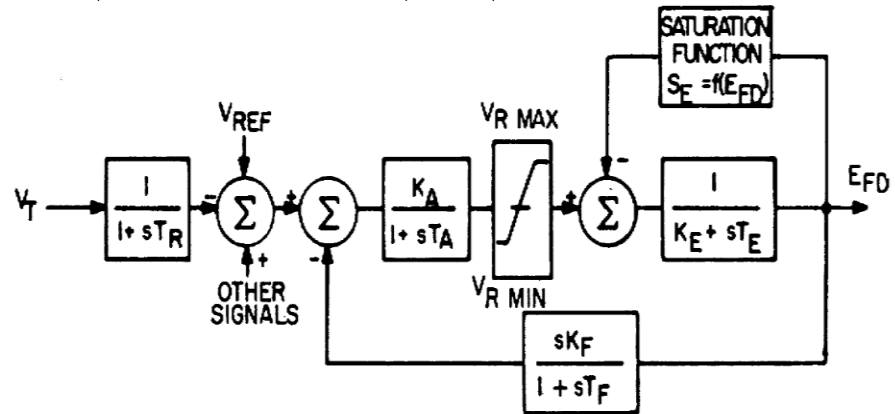
- Assume previous GENROU case with saturation. Then add a IEEE T1 exciter with  $K_a=50$ ,  $T_a=0.04$ ,  $K_e=-0.06$ ,  $T_e=0.6$ ,  $V_{r_{max}}=1.0$ ,  $V_{r_{min}}=-1.0$  For saturation assume  $Se(2.8) = 0.04$ ,  $Se(3.73)=0.33$

- Saturation function is  $0.1621(E_{fd}-2.303)^2$  (for  $E_{fd} > 2.303$ ); otherwise zero

- $E_{fd}$  is initially 3.22
- $Se(3.22)*E_{fd}=0.437$
- $(V_r-Se*E_{fd})/K_e=E_{fd}$

- $V_r = 0.244$

- $V_{ref} = 0.244/K_a + V_T = 0.0488 + 1.0946 = 1.09948$

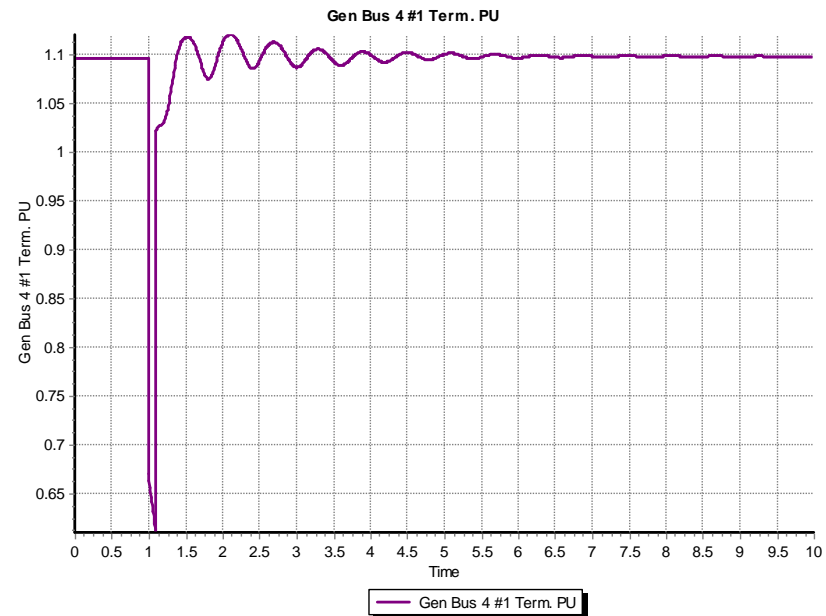
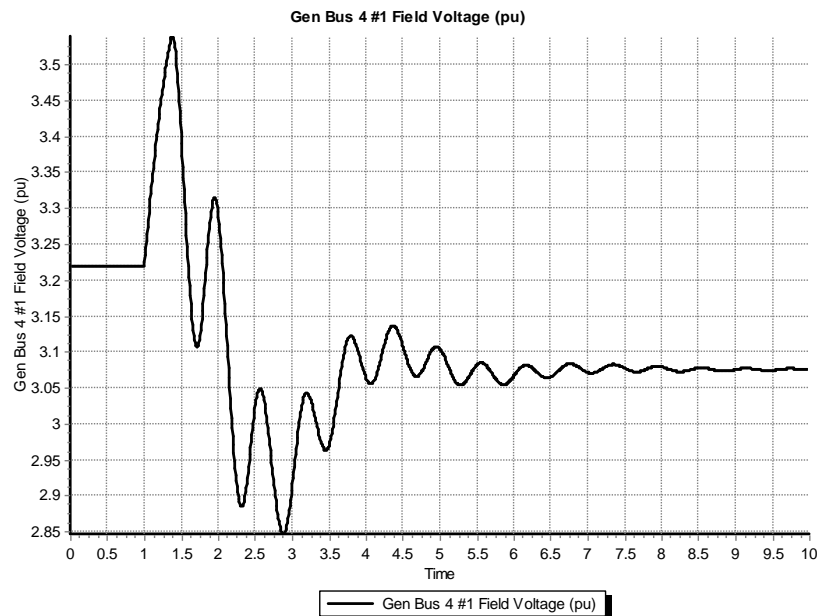


B4\_GENROU\_Sat\_IEEET1

# IEEE T1 Example



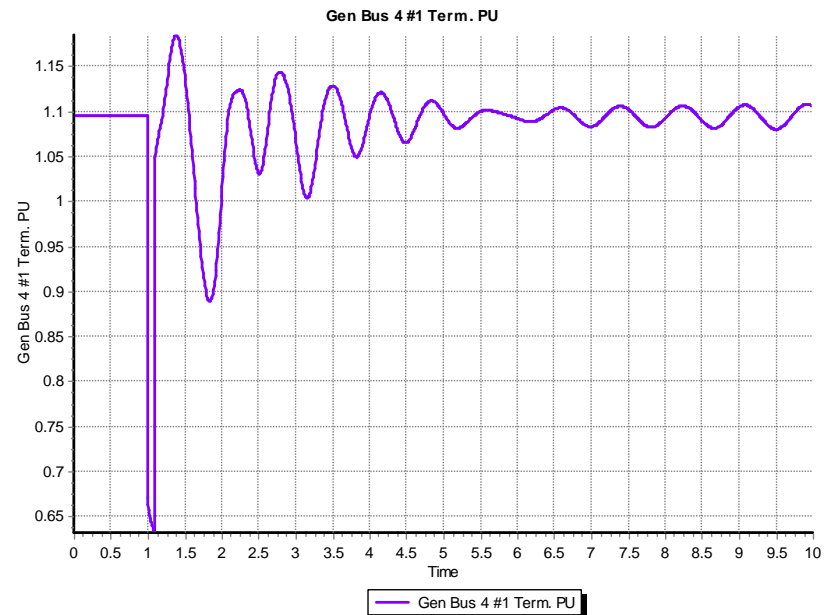
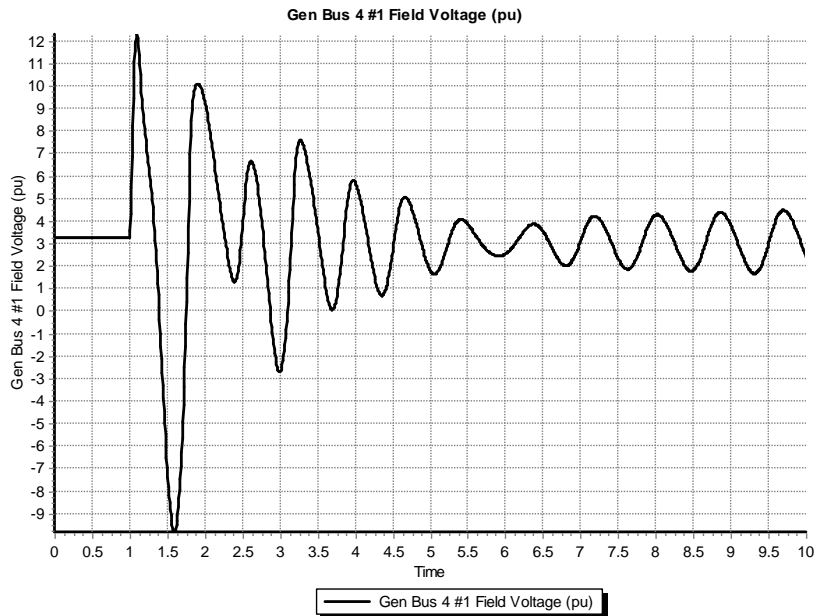
- For 0.1 second fault (from before), plot of Efd and the terminal voltage is given below
- Initial  $V_4=1.0946$ , final  $V_4=1.0973$ 
  - Steady-state error depends on the value of  $K_a$



# IEEE T1 Example



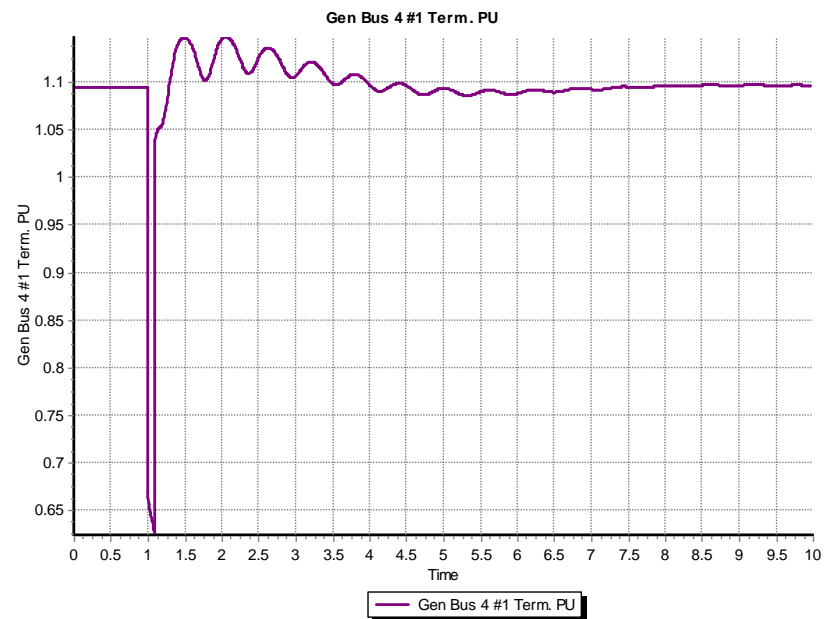
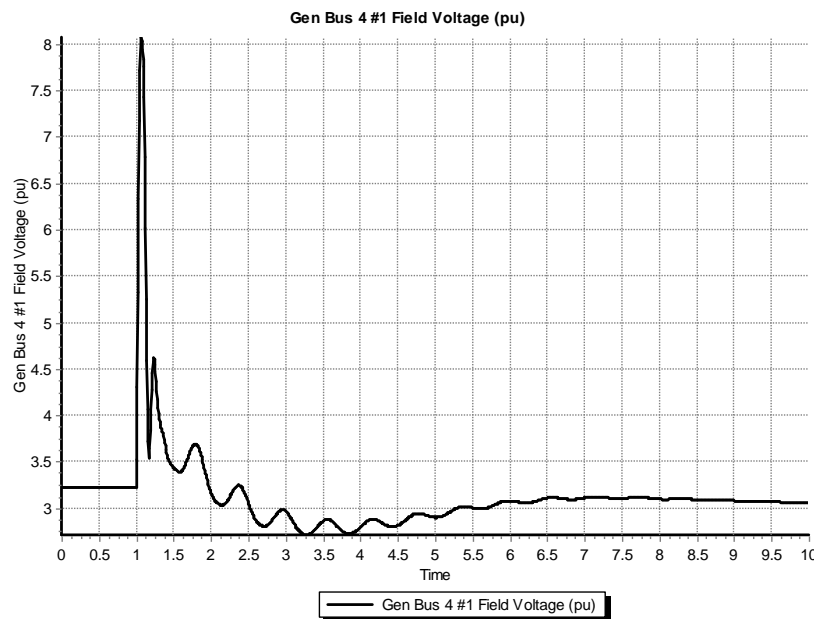
- Same case, except with  $K_a=500$  to decrease steady-state error, no  $V_r$  limits; this case is actually unstable



# IEEE T1 Example



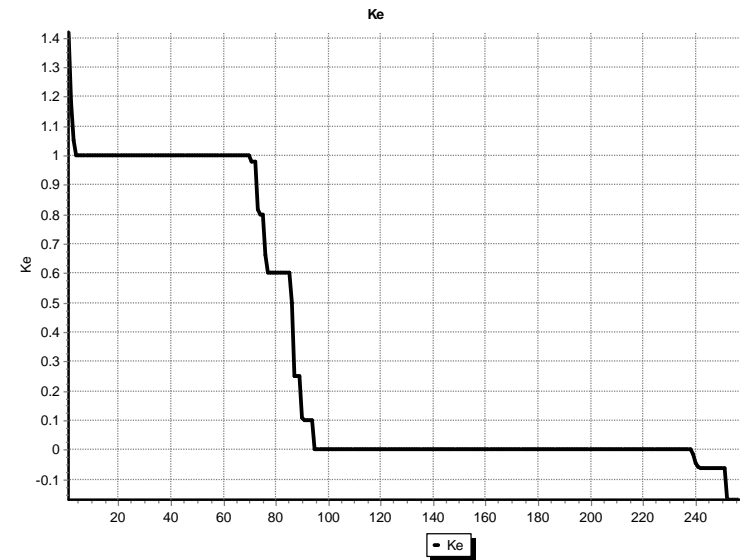
- With  $K_a=500$  and rate feedback,  $K_f=0.05$ ,  $T_f=0.5$
- Initial  $V_4=1.0946$ , final  $V_4=1.0957$



# WECC Case Type 1 Exciters

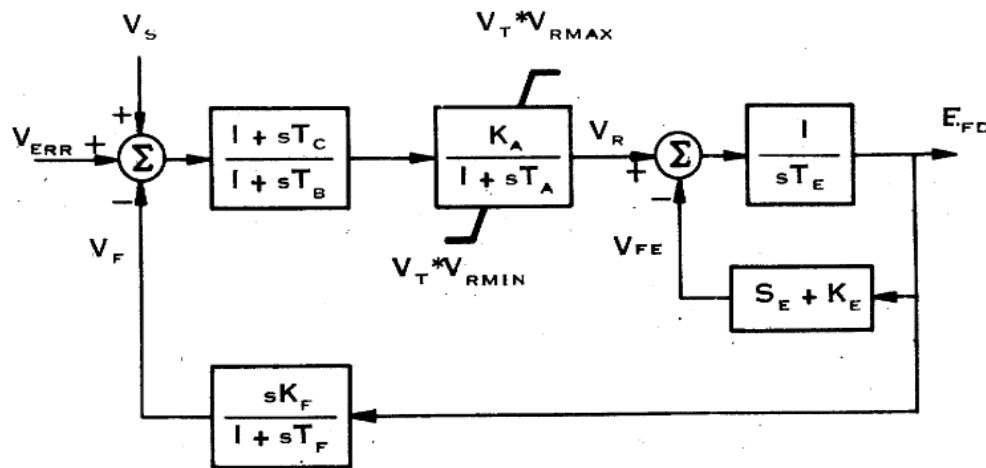


- In a recent WECC case with 2782 exciters, 58 are modeled with the IEEE T1, 257 with the EXDC1 and none with the ESDC1A
- Graph shows  $K_E$  value for the EXDC1 exciters in case; about 1/3 are separately excited, and the rest self excited
  - Value of  $K_E$  equal zero indicates code should set  $K_E$  so  $V_r$  initializes to zero; this is used to mimic the operator action of trimming this value



# DC2 Exciters

- Other dc exciters exist, such as the EXDC2, which is quite similar to the EXDC1; about 41 WECC exciters are of this type



Vr limits are multiplied by the terminal voltage

Fig. 4. Type DC2 - DC Commutator Exciter

Image Source: Fig 4 of "Excitation System Models for Power Stability Studies," IEEE Trans. Power App. and Syst., vol. PAS-100, pp. 494-509, February 1981

# ESDC4B



- Newer dc model introduced in 421.5-2005 in which a PID controller is added; might represent a retrofit

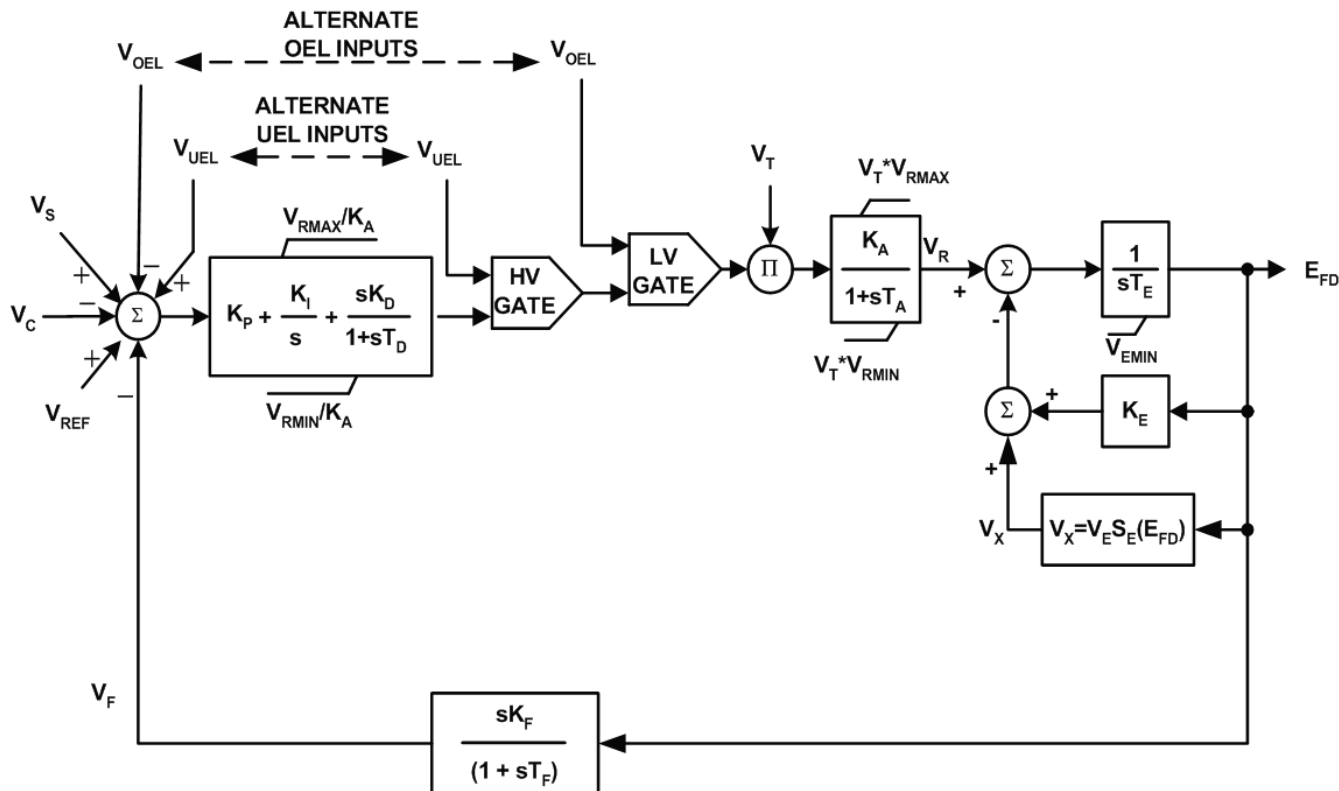


Image Source: Fig 5-4 of IEEE Std 421.5-2005



# Desired Performance



- A discussion of the desired performance of exciters is contained in IEEE Std. 421.2-2014 (update from 1990)
- Concerned with
  - large signal performance: large, often discrete change in the voltage such as due to a fault; nonlinearities are significant
    - Limits can play a significant role
  - small signal performance: small disturbances in which close to linear behavior can be assumed
- Increasingly exciters have inputs from power system stabilizers, so performance with these signals is important

# Transient Response

- Figure shows typical transient response performance to a step change in input

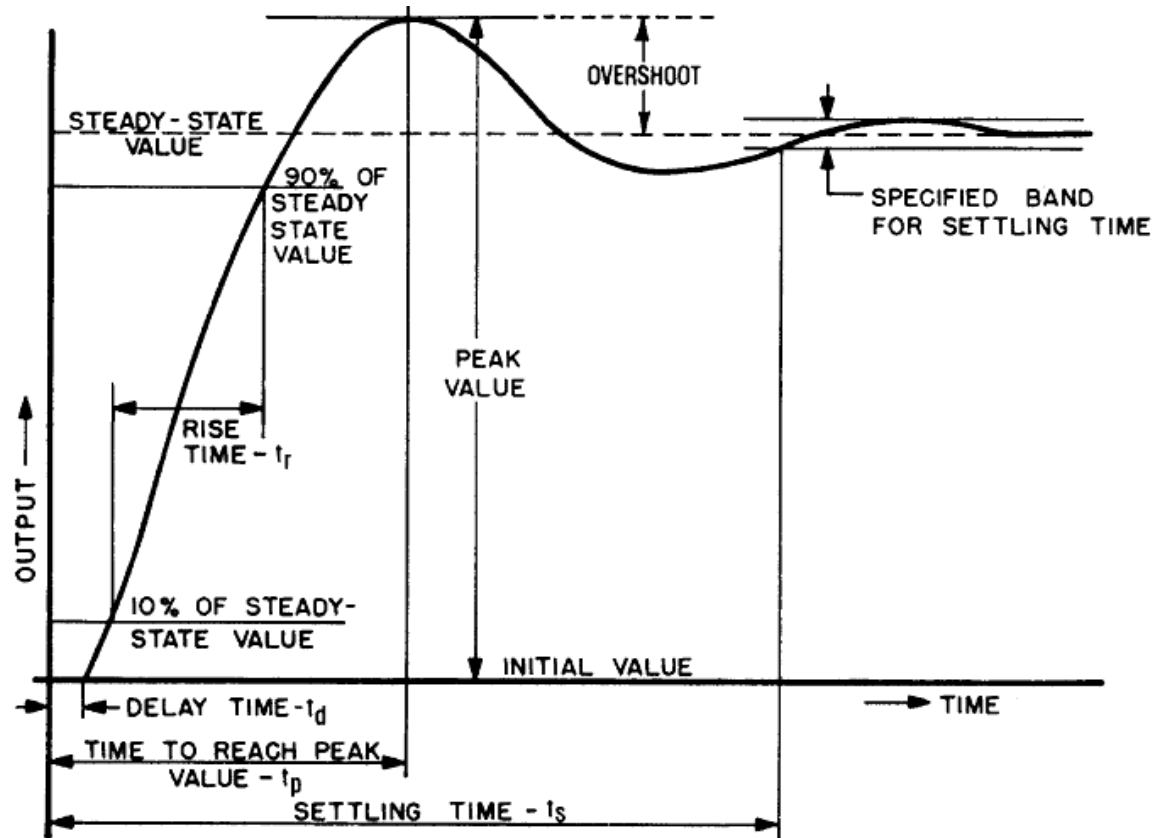


Image Source: IEEE Std 421.2-1990, Figure 3

# Small Signal Performance



- Small signal performance can be assessed by either the time responses, frequency response, or eigenvalue analysis
- Figure shows the typical open loop performance of an exciter and machine in the frequency domain

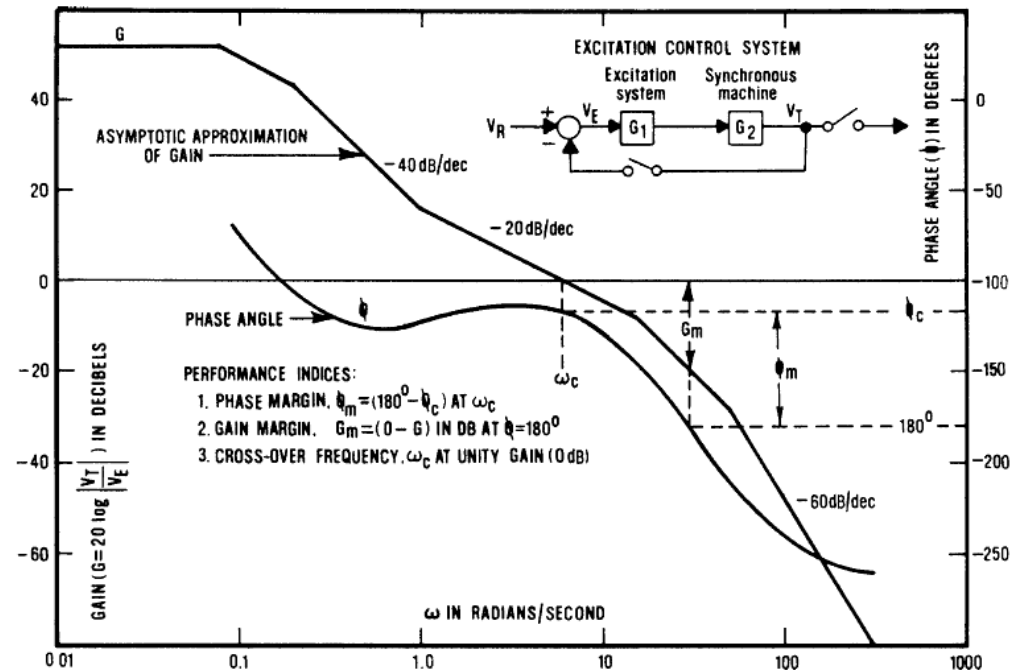
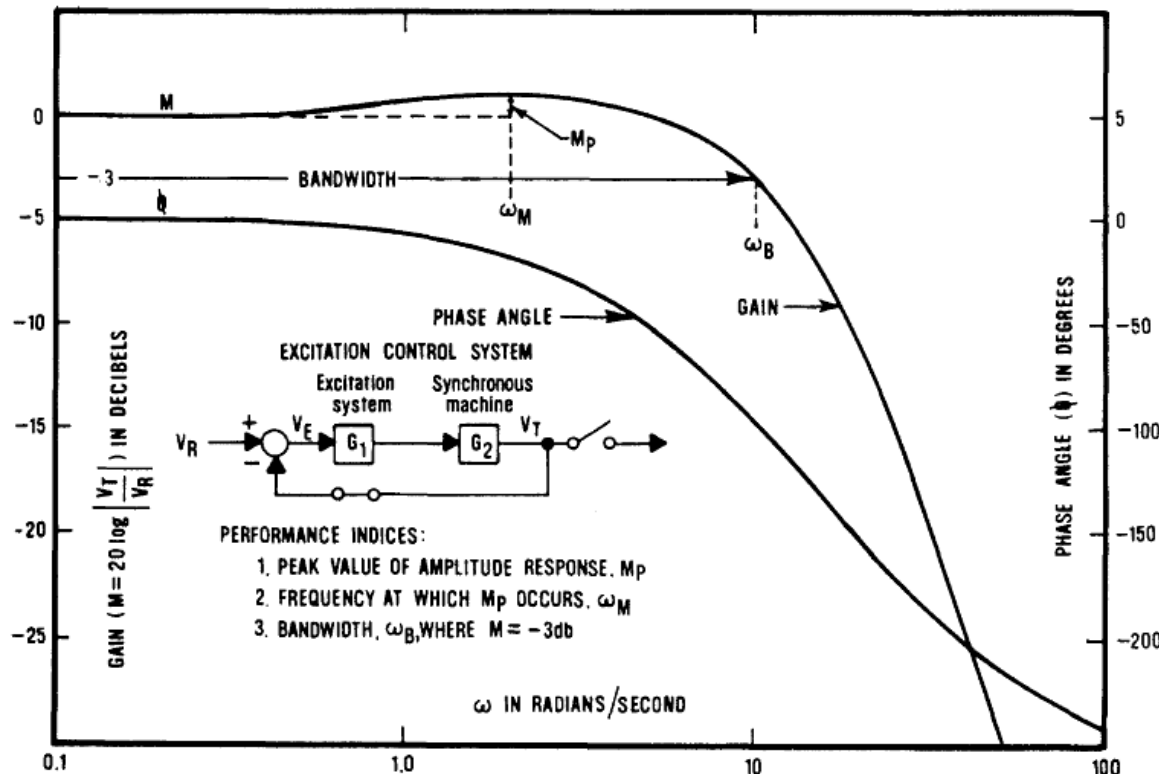


Figure 4—Typical Open-Loop Frequency Response of an Excitation Control System with the Synchronous Machine Open-Circuited

# Small Signal Performance



- Figure shows typical closed-loop performance



Peak value of  $M_p$  indicates relative stability; too large a value indicates overshoot

Note system connection is open

A larger bandwidth indicates a faster response

# AC Exciters



- Almost all new exciters use an ac source with an associated rectifier (either from a machine or static)
- AC exciters use an ac generator and either stationary or rotating rectifiers to produce the field current
  - In stationary systems the field current is provided through slip rings
  - In rotating systems since the rectifier is rotating there is no need for slip rings to provide the field current
  - Brushless systems avoid the anticipated problem of supplying high field current through brushes, but these problems have not really developed

# AC Exciter System Overview

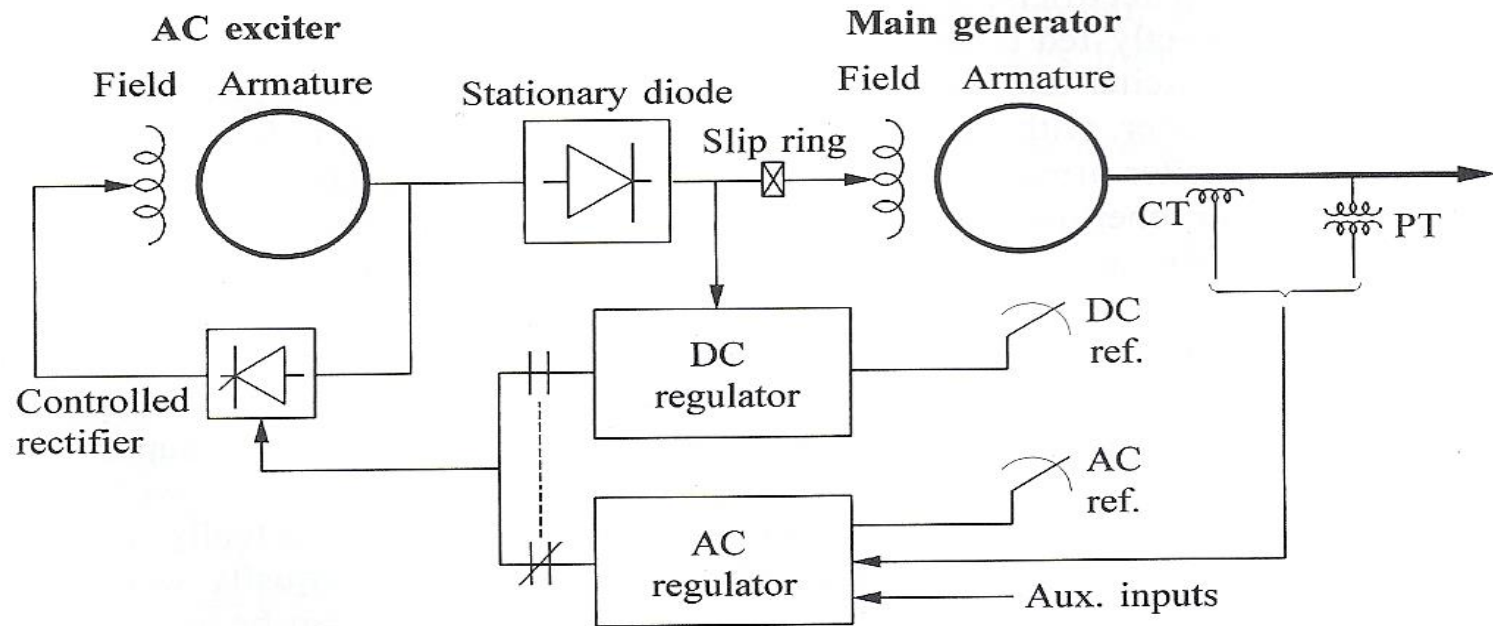


Figure 8.3 Field-controlled alternator rectifier excitation system

# ABB UNICITER



## UNICITER® Brushless Excitation Brushless excitation system – Electrical diagram

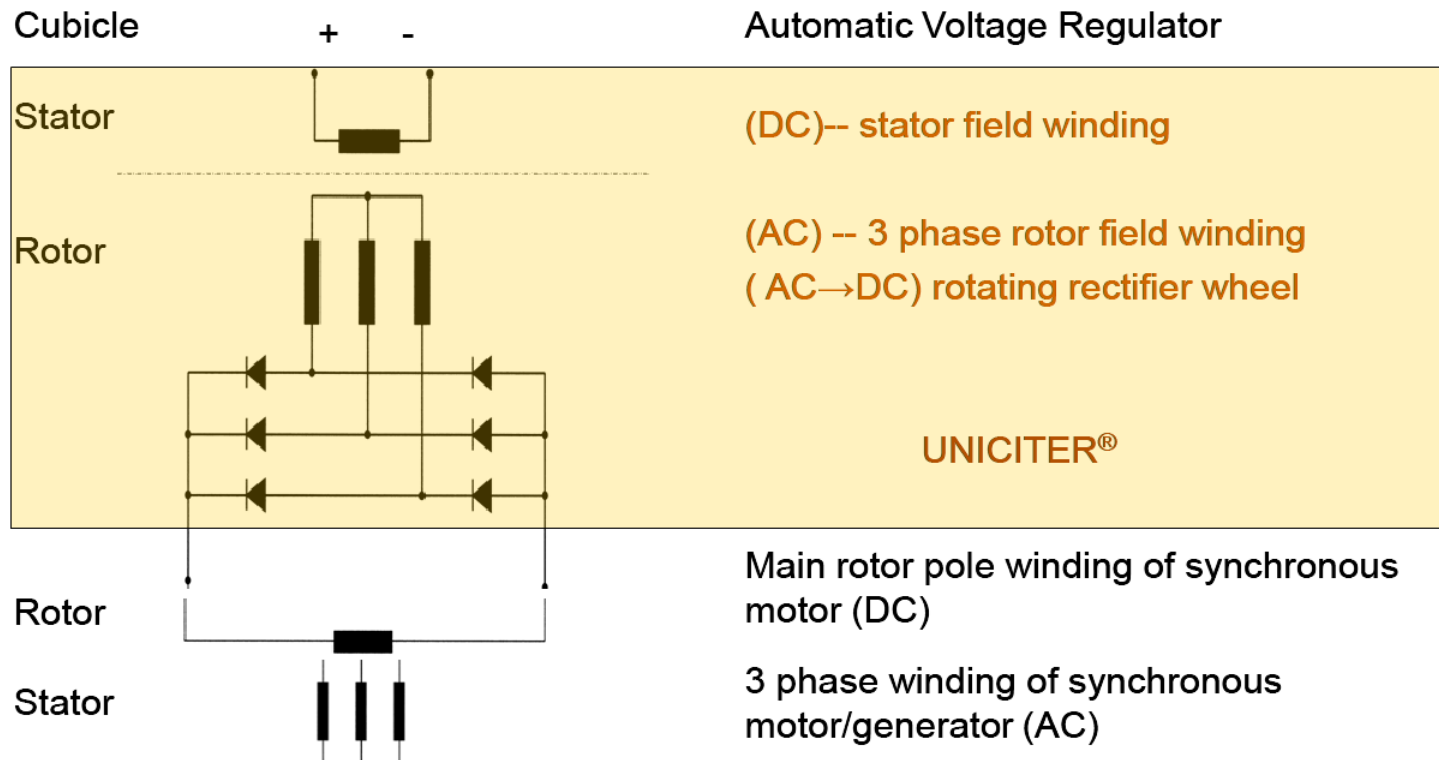


Image source: [www02.abb.com](http://www02.abb.com), Brushless Excitation Systems Upgrade,

# ABB UNICITER Example



## UNICITER® Example Hydro Power Plant – Horizontal - Switzerland



- Old DC commutator exciter by Brown Boveri
- Date of manufacture: 1960



New UNICITER® by ABB  
GTSC Birr

Image source: [www02.abb.com](http://www02.abb.com), Brushless Excitation Systems Upgrade,



# ABB UNICITER Rotor Field

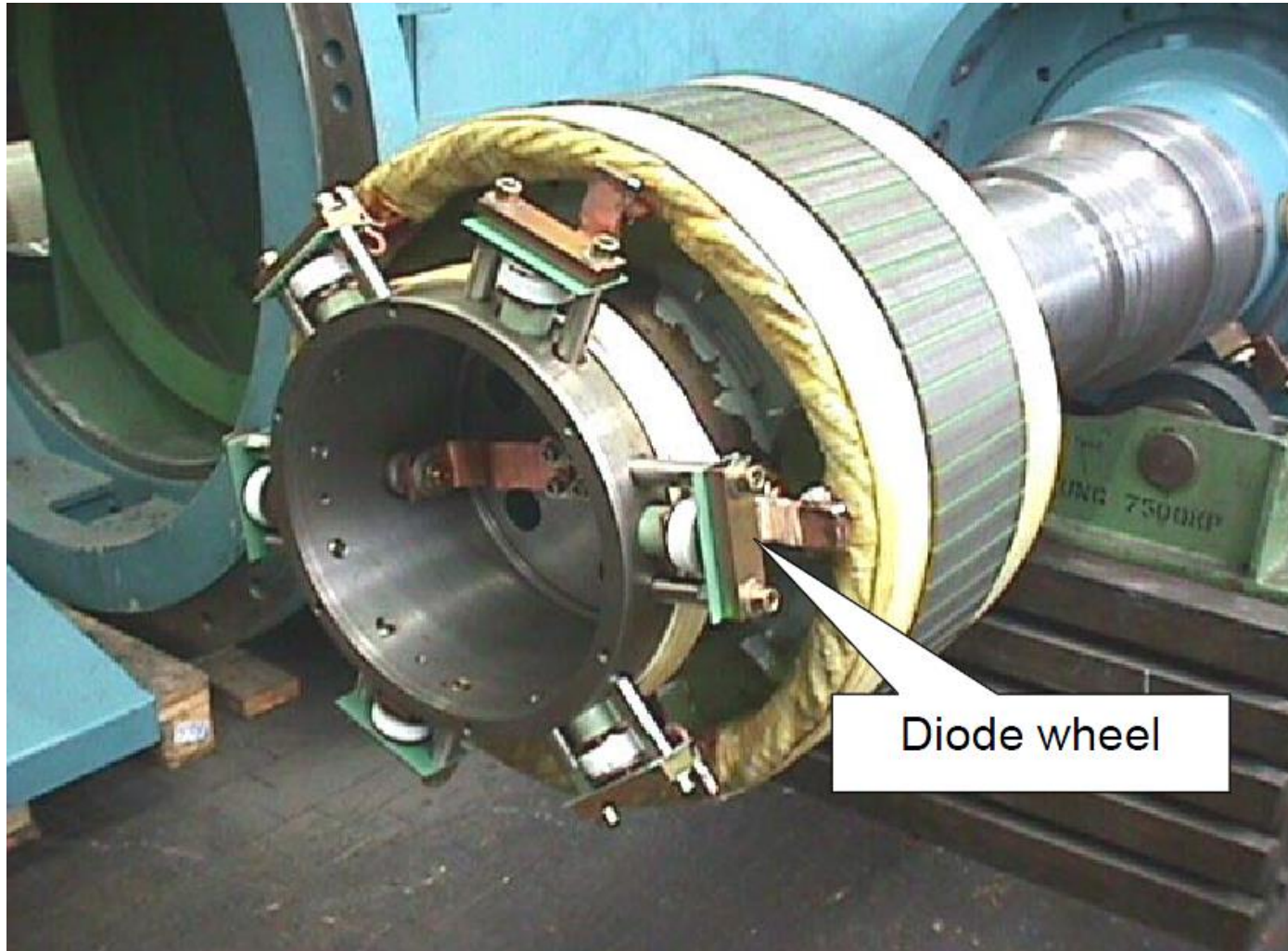
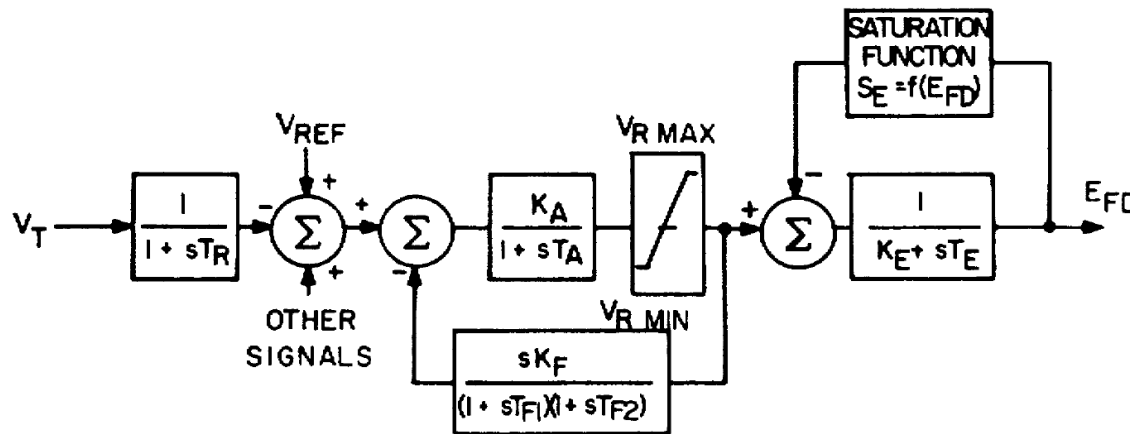


Image source: [www02.abb.com](http://www02.abb.com), Brushless Excitation Systems Upgrade,

# AC Exciter Modeling



- Originally represented by IEEE T2 shown below

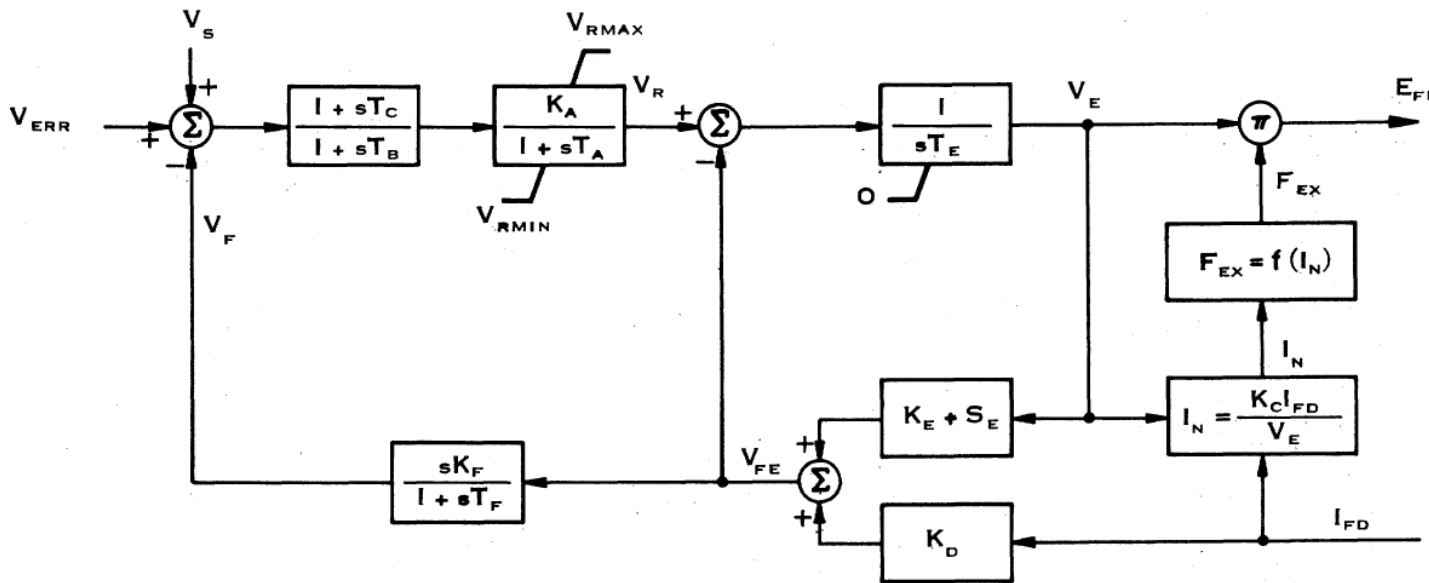


Exciter model is quite similar to IEEE T1

# EXAC1 Exciter



- The  $F_{EX}$  function represent the rectifier regulation, which results in a decrease in output voltage as the field current is increased



About 5% of WECC exciters are EXAC1

$K_D$  models the exciter machine reactance

Image Source: Fig 6 of "Excitation System Models for Power Stability Studies," IEEE Trans. Power App. and Syst., vol. PAS-100, pp. 494-509, February 1981

# EXAC1 Rectifier Regulation

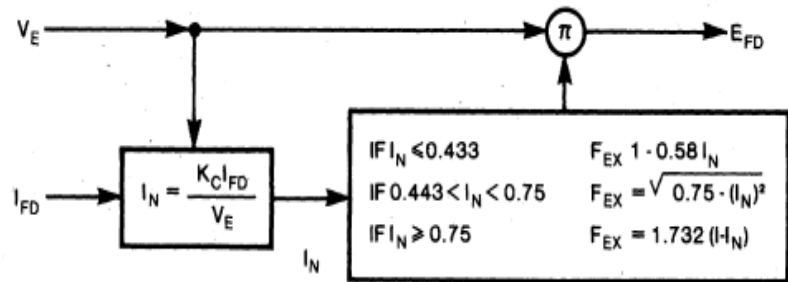
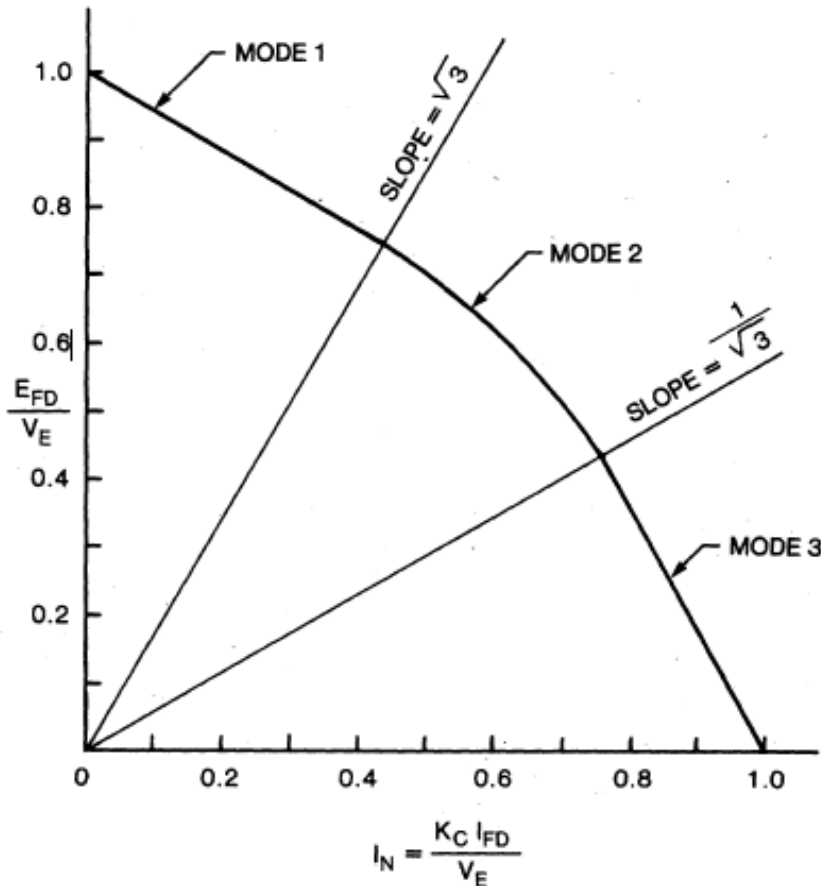


Fig. E.2. Rectifier Regulation Equations

$K_c$  represents the commuting reactance

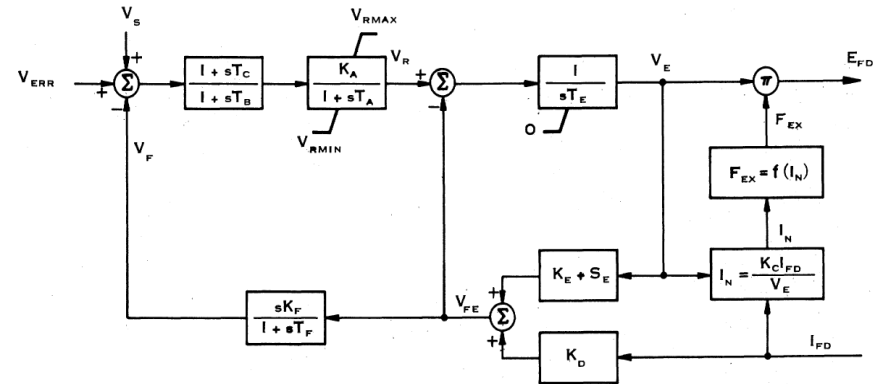
There are about 6 or 7 main types of ac exciter models

Image Source: Figures E.1 and E.2 of "Excitation System Models for Power Stability Studies," IEEE Trans. Power App. and Syst., vol. PAS-100, pp. 494-509, February 1981

# Initial State Determination, EXAC1



- To get initial states  $E_{fd}$  and  $I_{fd}$  would be known and equal
- Solve  $V_e * F_{ex}(I_{fd}, V_e) = E_{fd}$ 
  - Easy if  $K_c=0$ , then  $I_n=0$  and  $F_{ex} = 1$
  - Otherwise the  $F_{EX}$  function is represented by three piecewise functions; need to figure out the correct segment; for example for Mode 3



$$F_{ex} = \frac{E_{fd}}{V_e} = 1.732(I_{fd} - I_n) = 1.732 \left( 1 - \frac{K_c I_{fd}}{V_e} \right)$$

$$\text{Rewrite as } \frac{E_{fd}}{1.732} = V_e - K_c I_{fd} \rightarrow \frac{E_{fd}}{1.732} + K_c I_{fd}$$

Need to check to make sure we are on this segment

# Static Exciters

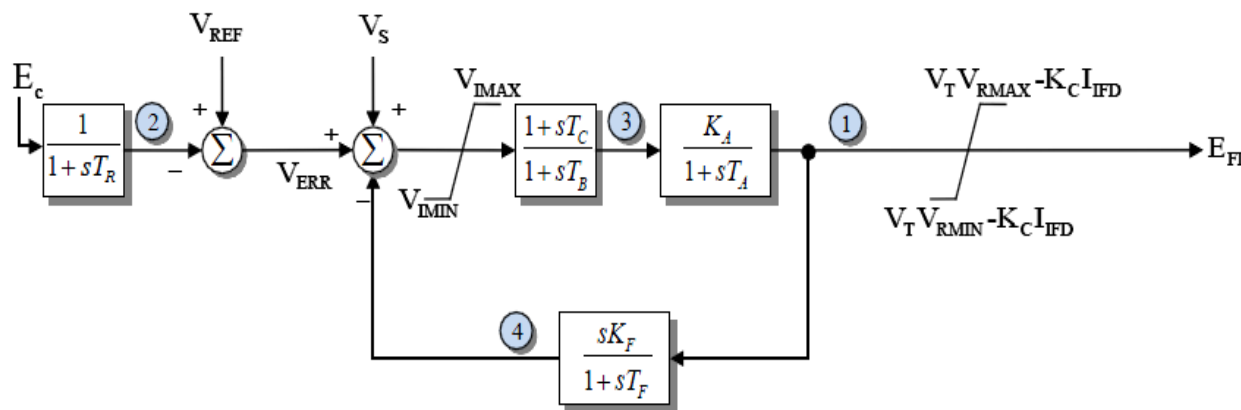


- In static exciters the field current is supplied from a three phase source that is rectified (i.e., there is no separate machine)
- Rectifier can be either controlled or uncontrolled
- Current is supplied through slip rings
- Response can be quite rapid

# EXST1 Block Diagram



- The EXST1 is intended to model rectifier in which the power is supplied by the generator's terminals via a transformer
  - Potential-source controlled-rectifier excitation system
- The exciter time constants are assumed to be so small they are not represented



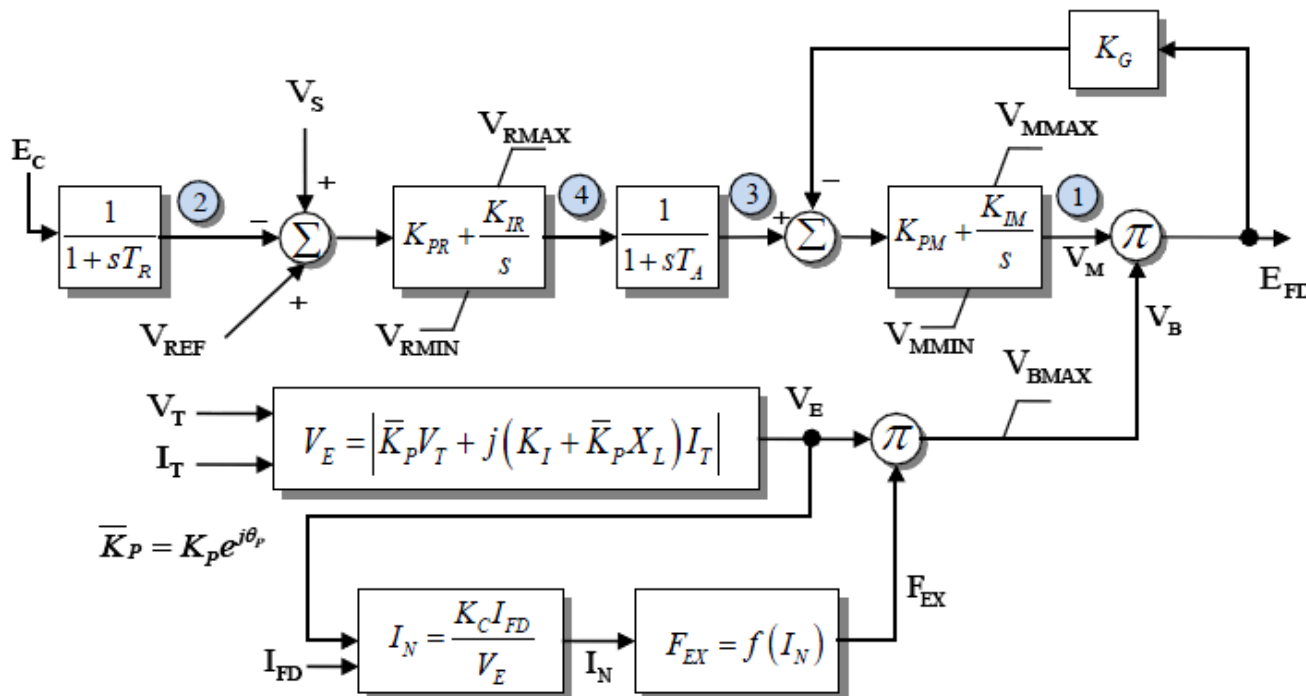
Most common exciter in WECC with about 29% modeled with this type

Kc represents the commuting reactance

# EXST4B



- EXST4B models a controlled rectifier design; field voltage loop is used to make output independent of supply voltage



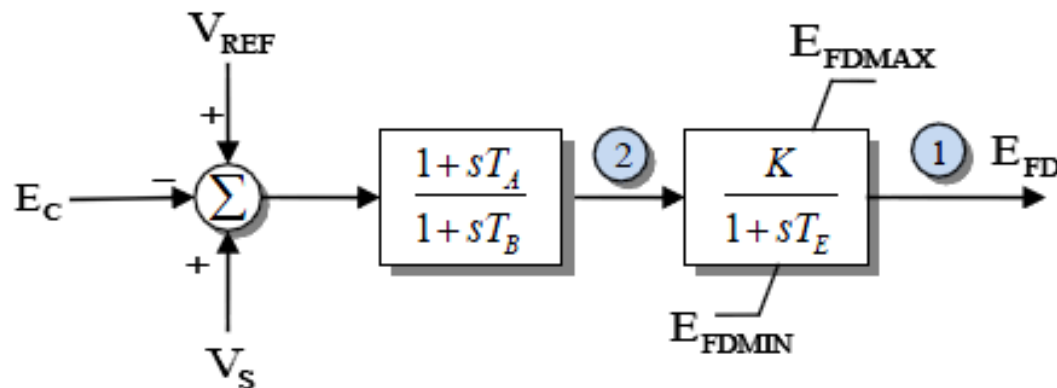
Second most common exciter in WECC with about 13% modeled with this type, though  $V_e$  is almost always independent of  $I_T$



# Simplified Excitation System Model



- A very simple model call Simplified EX System (SEXS) is available
  - Not now commonly used; also other, more detailed models, can match this behavior by setting various parameters to zero



# Compensation



- Often times it is useful to use a compensated voltage magnitude value as the input to the exciter
  - Compensated voltage depends on generator current; usually  $R_c$  is zero

$$E_c = \left| \bar{V}_t + (R_c + jX_c) I_T \right|$$

Sign convention is from IEEE 421.5

- PSLF and PowerWorld model compensation with the machine model using a minus sign
  - Specified on the machine base

$$E_c = \left| \bar{V}_t - (R_c + jX_c) I_T \right|$$

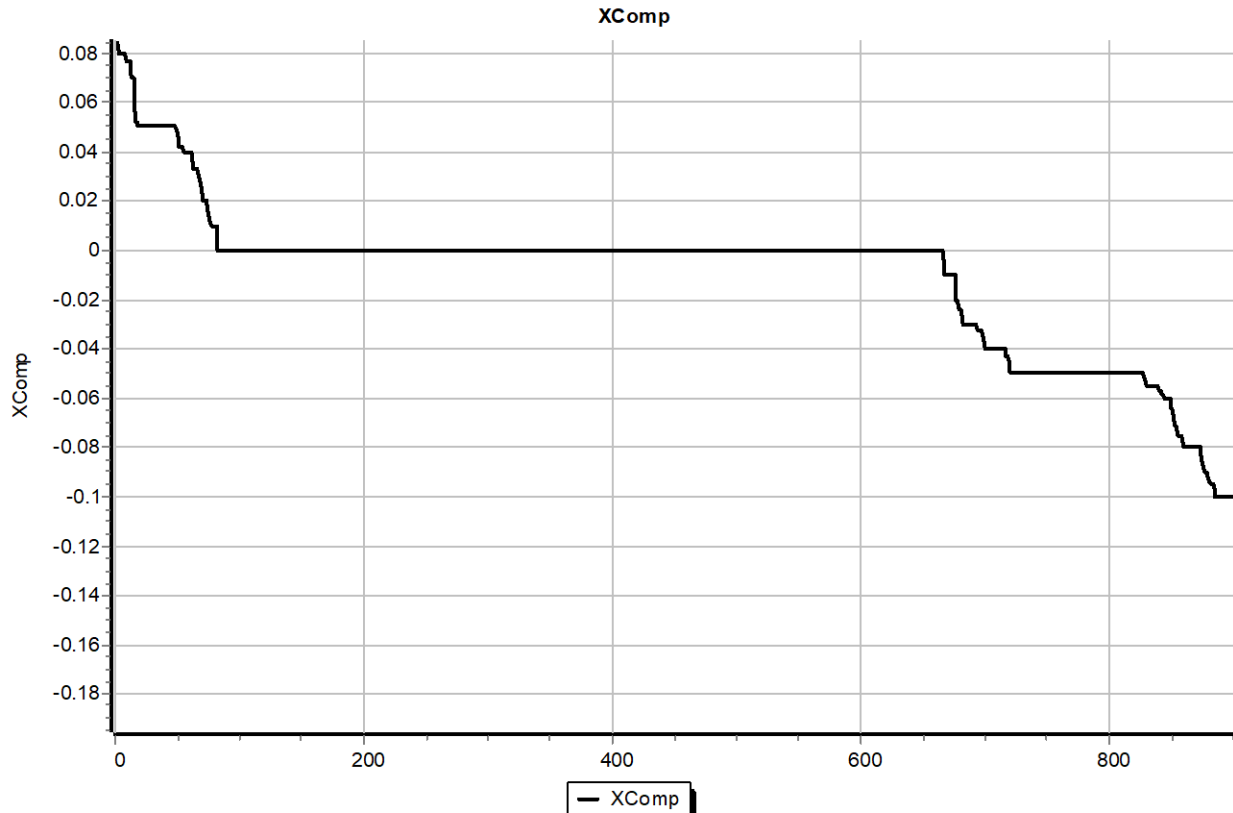
- PSSE requires a separate model with their COMP model also using a negative sign

# Compensation



- Using the negative sign convention
  - if  $X_c$  is negative then the compensated voltage is within the machine; this is known as droop compensation, which is used reactive power sharing among multiple generators at a bus
  - If  $X_c$  is positive then the compensated voltage is partially through the step-up transformer, allowing better voltage stability
    - A nice reference is C.W. Taylor, "Line drop compensation, high side voltage control, secondary voltage control – why not control a generator like a static var compensator," IEEE PES 2000 Summer Meeting

# Example Compensation Values



Negative values are within the machine

Graph shows example compensation values for large system; overall about 30% of models use compensation